

Combined fictitious-sources–scattering-matrix method

G erard Tayeb and Stefan Enoch

Institut Fresnel, Unit e Mixte de Recherche 6133, Centre National de la Recherche Scientifique, Facult e des Sciences et Techniques de Saint-J er me, Case 161, 13397 Marseille Cedex 20, France

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We describe a way to combine the method of fictitious sources and the scattering-matrix method. The resulting method presents concurrently the advantages of these two rigorous methods. It is able to solve efficiently electromagnetic problems in which the structure is made up of a jacket containing an arbitrary set of scatterers. The method is described in a two-dimensional case, but the basic ideas could be easily extended to three-dimensional cases.   2004 Optical Society of America

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1. INTRODUCTION

The scattering-matrix method (SMM)¹ applies to the scattering of a finite set of parallel cylinders of arbitrary cross section and arbitrary electromagnetic constants placed in a single homogeneous medium. It is based on the expansion of the fields in terms of Fourier–Bessel series around each cylinder. By using the scattering matrices of each cylinder and the translation properties of Fourier–Bessel functions, the method leads to the inversion of a linear set of equations. The method is rigorous and very efficient. It can deal for, instance, with photonic crystals (with or without defects) made up of several hundreds of cylinders on a standard personal computer.

On the other hand the SMM is not able to deal with certain configurations of interest, especially when the set of cylinders is surrounded by a jacket. In the case of a jacket of circular cross section, it is possible to extend the SMM by expanding the field around the jacket by using again Fourier–Bessel series.² This idea has also been implemented for the study of microstructured optical fibers.³

In this paper we extend the method to the case of a jacket of arbitrary cross section. For that purpose, we combine the SMM with the method of fictitious sources (MFS).

The MFS method⁴ is able to solve the problem of scattering from arbitrary scatterers. In this method, the space is divided into different regions where the field is represented as the field radiated by a set of fictitious sources with unknown intensities. These intensities are obtained by matching the fields at the boundaries of the regions by using a least-squares technique.

The basic idea of the method proposed in this paper is to use the SMM to build a set of functions that correctly represent the field inside the jacket. These functions are then used to solve a fictitious sources problem on the boundary of the jacket.

With this combined use of the SMM and MFS, we enjoy concurrently the advantages of these two rigorous methods and can address efficiently new classes of problems,

such as a finite dielectric body drilled by galleries, which can be, for instance, a macroporous silicon crystal.

The problem under study and our notation are described in Section 2. In Sections 3 and 4, we recall the basis of the MFS and the SMM. Section 5 describes the combined use of these two methods, and Section 6 provides some numerical illustrations of the resulting method.

2. SETTING OF THE PROBLEM

Throughout the paper, we use an orthogonal coordinate system with unit vectors \hat{x} , \hat{y} , \hat{z} . We consider time-harmonic problems, and the fields are represented by complex quantities with a time dependence in $\exp(-i\omega t)$. We denote by ϵ_0 and μ_0 the permittivity and the permeability, respectively, of vacuum and by $k_0 = 2\pi/\lambda_0 = \omega(\epsilon_0\mu_0)^{1/2}$ the wave number. For simplicity, we assume that all the media have the permeability μ_0 , but the principle of the method would remain unchanged if this were not the case. We consider a two-dimensional problem, assuming that the entire structure is invariant along the z axis (Fig. 1). The cylindrical scatterer is limited by its external boundary C_0 . The exterior of C_0 is the domain Ω_e filled with a medium of optical index n_e (n_e may be complex, and we note that $k_e = k_0 n_e$). The interior of C_0 is the domain Ω_i . It is filled with a medium of optical index n_i (gray region in Fig. 1; n_i may be complex, and we note that $k_i = k_0 n_i$), but it also contains cylindrical rods with boundaries C_j ($j = 1, 2, 3, \dots$) filled with arbitrary media.

The structure is illuminated by an incident field coming from the exterior (this assumption is made for clarity, but the method can also deal with an excitation from inside C_0 with very little modification). This incident field is also assumed to be z -independent. For instance, it can be a plane wave, or the field emitted by one (or several) line source(s) placed outside C_0 . It is well known that in that case the problem can be reduced to two independent problems: the s polarization case, where the electric field is

parallel to the z axis, and the p polarization case, where the magnetic field is parallel to the z axis. Each of these cases leads to a scalar problem where the unknown u is the z component of either \mathbf{E} or \mathbf{H} : $u = E_z$ (in s polarization) or $u = H_z$ (in p polarization). We denote by u^{inc} the incident field and by u^{scat} the scattered field in such a way that the total field is

$$\begin{cases} u = u^{inc} + u^{scat} & \text{in } \Omega_e \text{ (outside } C_0) \\ u = u^{int} & \text{in } \Omega_i \text{ (inside } C_0) \end{cases} \quad (1)$$

3. METHOD OF FICTITIOUS SOURCES

The MFS is a versatile and reliable method able to deal with many scattering problems. It relies on a simple idea: The electromagnetic field in the various domains of the diffracting structure is expressed as a combination of fields radiated by adequate electromagnetic sources. These sources have no physical existence, and this is why we have called them “fictitious” sources. They are located in homogeneous regions and not on the interfaces. In other words, one can consider that they generate electromagnetic fields that faithfully map the actual field, thus they form a convenient basis for this field. From a numerical point of view, proper bases are those capable of representing the solution with the fewest number of functions. Obviously, the quality of the bases is closely linked with the nature of the sources and their location. The freedom in the choice of the sources provides a great adaptability to various complex problems.

The MFS has been developed in our laboratory in the past decade, from both theoretical and numerical points of view.⁴⁻⁸ Almost at the same time and independently, some other groups have worked on the same basic ideas,⁹⁻¹⁴ but their approaches are slightly different from ours. In fact, one of the first attempts at using this method is probably due to Kupradze.¹⁵ The method has been developed and applied to a large collection of problems, and a good review can be found in Ref. 16. It is not our goal to depict here all the details of the MFS; it will be sufficient for our purpose to give an outline of the general principle.

Let us consider the same situation as in Fig. 1, but without the inclusions: The interior region Ω_i is thus filled with a homogeneous material of optical index n_i (Fig. 2). Let us imagine a set of N fictitious sources $S_{e,n}$ ($n = 1, 2, \dots, N$) located at N points $\mathbf{r}_{e,n}$ in Ω_i and assumed to radiate in free space filled with a medium of index n_e . Let us denote by $F_{e,n}(\mathbf{r})$ the field radiated by the

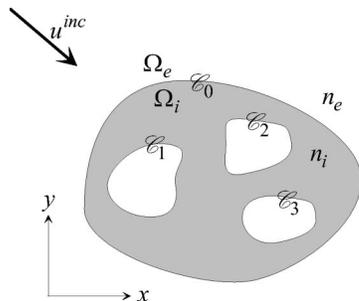


Fig. 1. Description of the problem.

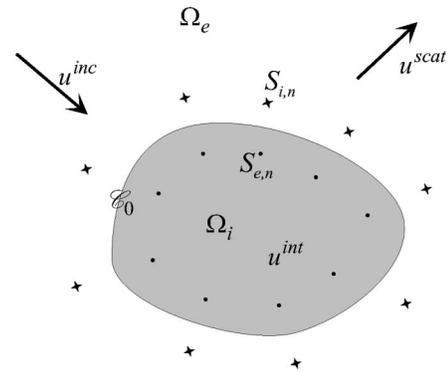


Fig. 2. Sources $S_{e,n}$ (represented by dots) radiate the fields $F_{e,n}(\mathbf{r})$ used to represent the scattered field u^{scat} in Ω_e ; sources $S_{i,n}$ (represented by stars) radiate the fields $F_{i,n}(\mathbf{r})$ used to represent the total field u^{int} in Ω_i .

source $S_{e,n}$. By construction, the fields $F_{e,n}(\mathbf{r})$ satisfy Maxwell’s equations in Ω_e and a radiation condition at infinity. A linear combination $\sum_n c_{e,n} F_{e,n}(\mathbf{r})$ can thus be regarded as an approximation $\tilde{u}^{scat}(\mathbf{r})$ of the diffracted field $u^{scat}(\mathbf{r})$ in Ω_e , where $c_{e,n}$ can be understood as the complex amplitude of the source $S_{e,n}$. So we obtain an approximation for the field in Ω_e as

$$\tilde{u}^{scat}(\mathbf{r}) \stackrel{\text{def}}{=} \sum_{n=1}^N c_{e,n} F_{e,n}(\mathbf{r}), \quad \text{in } \Omega_e, \quad (2)$$

and thus

$$u(\mathbf{r}) \approx u^{inc}(\mathbf{r}) + \sum_{n=1}^N c_{e,n} F_{e,n}(\mathbf{r}), \quad \text{in } \Omega_e. \quad (3)$$

In the same way, we imagine another set of fictitious sources $S_{i,n}$ ($n = 1, 2, \dots, N$) located at N points $\mathbf{r}_{i,n}$ in Ω_e and supposed to radiate in free space filled with a medium of index n_i . Let us denote by $F_{i,n}(\mathbf{r})$ the field radiated by the source $S_{i,n}$. The functions $F_{i,n}(\mathbf{r})$ verify Maxwell’s equations in Ω_i , and can be used to get an approximate expansion $\tilde{u}^{int}(\mathbf{r})$ of the total field $u^{int}(\mathbf{r})$ in Ω_i as

$$\tilde{u}^{int}(\mathbf{r}) \stackrel{\text{def}}{=} \sum_{n=1}^N c_{i,n} F_{i,n}(\mathbf{r}), \quad (4)$$

where $c_{i,n}$ can be understood as the complex amplitude of the source $S_{i,n}$.

Note that the nature of the sources can be chosen arbitrarily. In our case, we choose infinitely thin line sources parallel to the z axis: $S_{e,n}(\mathbf{r}) = 4i\delta(\mathbf{r} - \mathbf{r}_{e,n})$ and $S_{i,n}(\mathbf{r}) = 4i\delta(\mathbf{r} - \mathbf{r}_{i,n})$. In that case, they radiate the fields

$$F_{e,n}(\mathbf{r}) = H_0^{(1)}(k_e|\mathbf{r} - \mathbf{r}_{e,n}|), \quad (5)$$

$$F_{i,n}(\mathbf{r}) = H_0^{(1)}(k_i|\mathbf{r} - \mathbf{r}_{i,n}|). \quad (6)$$

Determination of the coefficients $c_{e,n}$ and $c_{i,n}$ is obtained by matching the boundary conditions on the cylinder surface C_0 . For a given function $\phi(\mathbf{r})$ let us denote by $D\phi(\mathbf{r})$ the value of its normal derivative on C_0 . The exact solution verifies

$$\begin{cases} u^{\text{inc}}(\mathbf{r}) + u^{\text{scat}}(\mathbf{r}) - u^{\text{int}}(\mathbf{r}) = 0 & \text{on } C_0 \\ Du^{\text{inc}}(\mathbf{r}) + Du^{\text{scat}}(\mathbf{r}) - pDu^{\text{int}}(\mathbf{r}) = 0 & \text{on } C_0' \end{cases} \quad (7)$$

where p is a polarization-dependent constant and

$$\begin{cases} p = 1 & \text{in the } s\text{-polarization case} \\ p = (n_e/n_i)^2 & \text{in the } p\text{-polarization case} \end{cases} \quad (8)$$

The coefficients $c_{e,n}$ and $c_{i,n}$ that give the better approximation for the fields of Eqs. (2) and (4) are those that match the boundary conditions in the best way. They are obtained by minimizing the two expressions derived from Eqs. (7) and defined on C_0 :

$$\begin{cases} u^{\text{inc}}(\mathbf{r}) + \sum_{n=1}^N c_{e,n} F_{e,n}(\mathbf{r}) - \sum_{n=1}^N c_{i,n} F_{i,n}(\mathbf{r}) \\ Du^{\text{inc}}(\mathbf{r}) + \sum_{n=1}^N c_{e,n} DF_{e,n}(\mathbf{r}) - p \sum_{n=1}^N c_{i,n} DF_{i,n}(\mathbf{r}) \end{cases} \quad (9)$$

This can be done in several ways. The simplest one is to use a point-matching method that enforces the vanishing of these two expressions on sample points on C_0 . In this way, we can get a system of $2N$ equations for the $2N$ unknowns $c_{e,n}$ and $c_{i,n}$. From our numerical experiments, it emerges that it is preferable to use an overdetermined system and to solve it in the least-squares sense.¹⁷ In this way, and for a given computation time, we obtain a better approximation for u^{scat} and u^{int} .

Of course, the efficiency of the method depends on the location and the number N of the fictitious sources. It can be shown that the precision of the method is related to the least-squares remainder obtained in the last step, which thus can be used to quantize the quality of the solution. This feature is quite helpful in the numerical implementation.

The interested reader will find more details in Ref. 4. We have also developed some “tricks” in order to place the

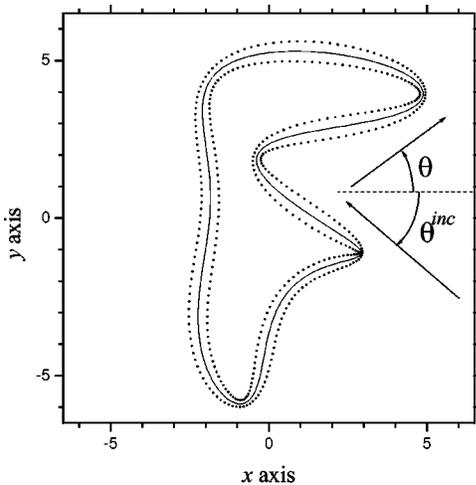


Fig. 3. Cross section of the cylinder and the two sets of sources. The profile is given by Eq. (10) and the values $c_{-5} = -0.1134 + i0.1310$, $c_{-4} = -0.0297 - i0.3238$, $c_{-3} = -0.4117 - i0.0973$, $c_{-2} = -0.1260 + i1.4149$, $c_{-1} = -2.3936 + i2.4031$, $c_0 = 0.5714 + i0.5000$, $c_1 = 1.5568 + i0.1876$, $c_2 = -0.1212 - i0.0197$, $c_3 = -0.8158 + i0.2155$, $c_4 = 0.2772 + i0.1039$, and $c_5 = -0.1532 - i0.0102$.

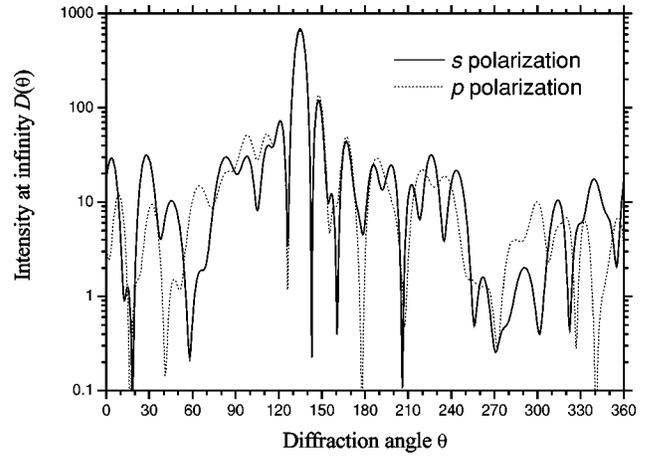


Fig. 4. Scattered intensity at infinity for both polarizations.

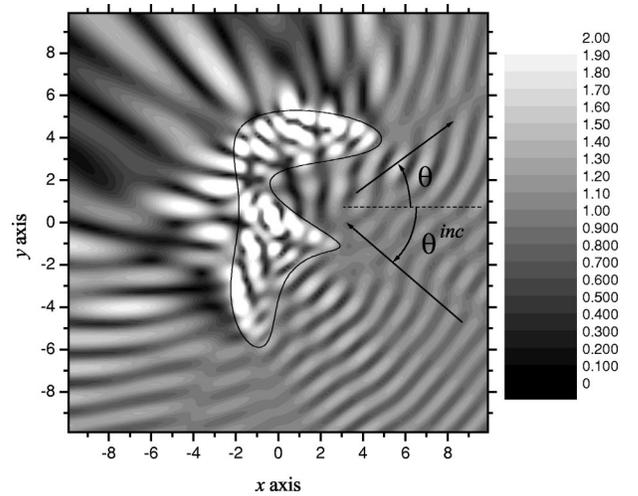


Fig. 5. Modulus of the total field in p polarization.

sources automatically. The general idea is to increase the number of sources where the radius of curvature of C_0 is lower. An example is given in Fig. 3. The cross section of the cylinder C_0 mimics a rounded letter F (first letter of Fresnel Institute), and it is given by the parametric equation

$$z(t) = x(t) + iy(t) = \sum_{n=-5,5} c_n \exp(in2\pi t). \quad (10)$$

The values of the coefficients c_n are given in the caption of Fig. 3. We use $N = 200$ sources in each region Ω_e and Ω_i and $2N$ sample points on C_0 . This cylinder of index $n_i = 1.5$ lies in vacuum ($n_e = 1$) and is illuminated at an incidence $\theta^{\text{inc}} = -45^\circ$ by a plane wave with wavelength $\lambda_0 = 2$ and unit amplitude. Figure 4 gives the intensity $D(\theta)$ scattered at infinity in the direction θ . It is defined in the following way: As a result of the asymptotic behavior of the Hankel function, the scattered field at infinity is written (with $r = |\mathbf{r}|$) as

$$u^{\text{scat}}(\mathbf{r}) \approx g(\theta) \frac{\exp(ik_e r)}{\sqrt{r}}, \quad (11)$$

and the intensity scattered at infinity is

$$D(\theta) = 2\pi|g(\theta)|^2. \quad (12)$$

Finally, Fig. 5 shows the field map in the vicinity of the scatterer.

4. THE SCATTERING-MATRIX METHOD

The SMM is able to solve the problem of the diffraction by an arbitrary set of parallel cylinders placed in a homogeneous medium. A detailed description of the problem can be found in Ref. 1. Note that the method is also called “multipole method” by other authors and has been used to study the local density of states in photonic crystals¹⁸ as well as microstructured optical fibers.³ We give below only an outline of the basic ideas.

We consider a set of N_c parallel cylinders C_j , as shown in Fig. 6. To lay the groundwork for Section 5 we assume that the medium outside the cylinders has the index n_i as defined in Section 2. The incident field u^{inc} can be arbitrary (plane wave, line source, etc.).

For any cylinder C_j , we consider a circle \mathcal{D}_j with center O_j in such a way that the cylinder is completely inside \mathcal{D}_j (Fig. 7). Because of the properties of the Helmholtz equation, the total field $u(P)$ at a point P on \mathcal{D}_j can be written as a Fourier–Bessel expansion. Denoting by $r_j(P)$ and $\theta_j(P)$ the polar coordinates in the local system (O_j, x_j, y_j) , we can write

$$u(P) = \sum_{m=-\infty, +\infty} [a_{j,m} J_m(k_i r_j(P)) + b_{j,m} H_m^{(1)}(k_i r_j(P))] \exp[im\theta_j(P)]. \quad (13)$$

The two terms in the preceding series of Eq. (13) can be interpreted in the following way. The second term satisfies a radiation condition and thus represents the field

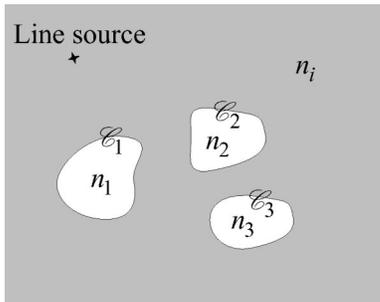


Fig. 6. Set of $N_c = 3$ parallel cylinders in a medium with index n_i in the case in which the incident field is created by a line source.

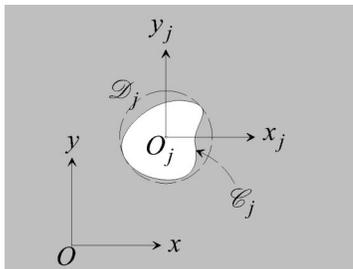


Fig. 7. Circle \mathcal{D}_j surrounding the cylinder C_j and local coordinate system.

scattered by the cylinder C_j . For each cylinder, this scattered field will be characterized by the matrix column \mathbf{b}_j containing the $b_{j,m}$ elements. The first term represents the local incident field on the cylinder C_j generated by the actual incident field u^{inc} as well as by the fields scattered by all the other cylinders C_k with $k \neq j$.

If we denote by \mathbf{a}_j the matrix column containing the $a_{j,m}$ and use translation properties of Bessel functions (Graf’s formula¹⁹), we can obtain¹ for any cylinder C_j a linear relationship,

$$\mathbf{a}_j = \mathbf{Q}_j + \sum_{k \neq j} \mathbf{T}_{j,k} \mathbf{b}_k, \quad (14)$$

where \mathbf{Q}_j is a known column matrix that represents the actual incident field on the cylinder C_j , and $\mathbf{T}_{j,k}$ is a known square matrix (its elements simply contain exponentials and Hankel functions).

For any cylinder C_j , another relation between \mathbf{b}_j and \mathbf{a}_j is provided by the scattering matrix \mathbf{S}_j of the cylinder. The diffracted field is linked to the local incident field by

$$\mathbf{b}_j = \mathbf{S}_j \mathbf{a}_j. \quad (15)$$

By eliminating \mathbf{a}_j from Eqs. (14) and (15) and then collecting the equations written for each cylinder, we obtain a linear system that gives the solution \mathbf{b}_j as

$$\begin{bmatrix} \mathbf{I} & -\mathbf{S}_1 \mathbf{T}_{1,2} & \cdots & -\mathbf{S}_1 \mathbf{T}_{1,N} \\ -\mathbf{S}_2 \mathbf{T}_{2,1} & \mathbf{I} & \cdots & -\mathbf{S}_2 \mathbf{T}_{2,N} \\ \vdots & \vdots & \ddots & \vdots \\ -\mathbf{S}_N \mathbf{T}_{N,1} & -\mathbf{S}_N \mathbf{T}_{N,2} & \cdots & \mathbf{I} \end{bmatrix} \begin{pmatrix} \mathbf{b}_1 \\ \mathbf{b}_2 \\ \vdots \\ \mathbf{b}_N \end{pmatrix} = \begin{pmatrix} \mathbf{S}_1 \mathbf{Q}_1 \\ \mathbf{S}_2 \mathbf{Q}_2 \\ \vdots \\ \mathbf{S}_N \mathbf{Q}_N \end{pmatrix}, \quad (16)$$

where \mathbf{I} denotes the identity matrix. For brevity, this equation will be written as

$$\mathbf{S}^{-1} \mathbf{B} = \mathbf{A} \quad (17)$$

and formally inverted as

$$\mathbf{B} = \mathbf{S} \mathbf{A}. \quad (18)$$

Now let us point out some features of the method.

1. In Eq. (18), column \mathbf{A} linearly depends on the actual incident field u^{inc} , and column \mathbf{B} contains the information on the field diffracted by the entire set of cylinders. In that sense \mathbf{S} is the scattering matrix of the set of cylinders. From Eq. (16), it appears that \mathbf{S}^{-1} is simply built as soon as the scattering matrices \mathbf{S}_j of all cylinders are known. This feature is interesting from a numerical point of view. It means that the individual scattering matrices \mathbf{S}_j can be constructed independently from the main code dealing with the set of cylinders. That is why we call this the scattering-matrix method. When the cylinders are circular, matrices \mathbf{S}_j are very simple and reduce to diagonal matrices whose elements can be expressed in closed form. When the cylinders are not circular, we use an external integral code to compute the \mathbf{S}_j . Another important point is that, when all the cylinders in the set are identical, all the matrices \mathbf{S}_j are also

identical (because they are defined in the local coordinate system centered on each cylinder).

2. Solving the system of Eq. (16) gives the $b_{j,m}$. From that knowledge, the total field u is given outside the circles \mathcal{D}_j by

$$u(P) = u^{\text{inc}}(P) + \sum_{j=1}^{N_c} \sum_{m=-\infty}^{+\infty} b_{j,m} H_m^{(1)}(k_i r_j(P)) \exp[im \theta_j(P)]. \quad (19)$$

The consequence is that everywhere outside the circles \mathcal{D}_j , the field u and also its derivatives are known in closed form by Eq. (19) and its derivatives.

3. For numerical purposes it is clear that the series in Eq. (13) has to be truncated. It can be shown that, as a result of properties of the Helmholtz equation, the terms of the series are decreasing extremely fast after a given threshold is reached. Assuming that we keep M terms in the series, columns \mathbf{a}_j and \mathbf{b}_j reduce to M elements, matrices \mathbf{S}_j have a rank M , and the linear system of Eq. (16) has a rank $N_c \times M$. In fact, the value of M is closely linked with the radius ρ of the cylinders (or of \mathcal{D}_j for non-circular cylinders) and the wavelength $\lambda_i = 2\pi/k_i$. A convenient value for M is given by the empirical rule $M \approx 40\rho/\lambda_i$ (taking for M an odd integer). With such value of M , the accuracy is better than 1%. As an example, for typical values used in photonic crystal problems, we generally take $M = 7$. This means that for $N = 100$ cylinders, we solve a 700×700 system.

4. The incident field is arbitrary. It can be a plane wave, a Gaussian beam, or the field emitted by one (or several) infinitely thin line source(s) parallel to the z axis acting as an antenna. For the purpose of this paper, we are mainly interested in this last case, and u^{inc} is then given by expressions similar to Eq. (6). Note that the incident field appears only in the second member of Eq. (16), and consequently, dealing with several incident fields is numerically very efficient.

5. Last, we must point out one limitation of the method. The Fourier-Bessel expansion in Eq. (13) is valid only if the circle \mathcal{D}_j lies in a homogeneous medium. This means that the circle that contains one cylinder cannot intersect the boundary of another cylinder. In other words, one can remember that the circles \mathcal{D}_j must have no intersections. Of course, this is always so when the cylinders are circular. In fact, for noncircular cylinders, the problem is much more subtle, and the method should also work in some cases where the circles intersect. This problem is similar to the problem of validity of the Rayleigh hypothesis in grating theory.²⁰

5. HYBRID METHOD COMBINING THE METHOD OF FICTITIOUS SOURCES AND THE SCATTERING-MATRIX METHOD

Let us come back to the original problem described in Section 2. This problem can be solved by a slight modification of the MFS described in Section 3.

Indeed, the scattered field can be still expressed as in Eq. (2) by using the same fictitious sources $S_{e,n}$ [line

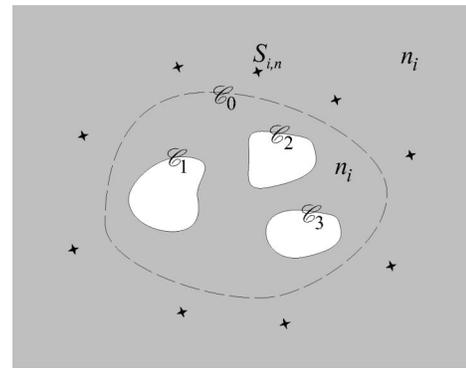


Fig. 8. Setting of the problem to obtain the functions $F_{i,n}(\mathbf{r})$.

sources that radiate $F_{e,n}(\mathbf{r})$ fields expressed as Hankel functions exactly as in Eq. (5)]. But in that case the $F_{i,n}(\mathbf{r})$ functions used to expand the field in Ω_i (inside C_0) must be changed. Let us consider the problem depicted in Fig. 8. The inclusions C_j are immersed in a medium with index n_i (the boundary C_0 of the external scatterer is suppressed). We keep the same line sources $S_{i,n}$ as in Section 3. The new $F_{i,n}(\mathbf{r})$ function is the total field when the structure of Fig. 8 is excited by the source $S_{i,n}$. By solving this problem as proposed in Section 4, $F_{i,n}(\mathbf{r})$ can be expressed by expansion (19). In other words $F_{i,n}(\mathbf{r})$ ($n = 1, 2, \dots, N$) is a set of solutions for the total field inside C_0 that are available in closed form and that can be used to expand the field in Ω_i following Eq. (4). Using expansion (19), we can compute the value of $F_{i,n}(\mathbf{r})$ and its normal derivative on C_0 and get expressions (9) to be minimized. This minimization gives the coefficients $c_{e,n}$ and $c_{i,n}$, and we finally get the expressions of the total field in closed form everywhere by using relation (3) in Ω_e and relation (4) in Ω_i .

6. NUMERICAL EXAMPLES

We begin our numerical examples by a comparison with the result of Ref. 2. With the notation of Section 2, C_0 is a circle of radius 10 (arbitrary units), the outside medium is vacuum ($n_e = 1$), and $n_i = 4$. The inclusions are circular cylinders of radius 0.8 filled with vacuum and arranged with a hexagonal symmetry, with the distance between the centers of the cylinders equal to 4. The central cylinder is suppressed. The structure is illuminated by a plane wave in p polarization, coming from the top of the figure, with amplitude normalized to unity. The wavelength $\lambda_0 = 22$ is chosen to get a resonant localized mode inside the structure. Figure 9 shows the modulus of the total field. It is quite similar to Fig. 7 of Ref. 2 except inside the small cylinders, where we are inclined to trust our computation. Note that the method described in Ref. 2 can deal with an external boundary C_0 with a circular shape only.

In a second example, we illustrate some possibilities of the method in a more complex situation. As in Section 3 the cross section of the cylinder C_0 is the letter F, but now with sharp edges (Fig. 10). The reason for this choice is only to prove that the method also works well in such a case, which is more difficult to solve than a cylinder with rounded boundaries. All the coordinates of C_0 corners in

the (x, y) plane have integer values that can be deduced from Fig. 10. The profile C_0 is described by a series identical to that in Eq. (10), but in this case we use a large number of c_n coefficients ($n = -100, 100$) to get a quasipolygonal shape. For the interested reader, some more details on the technical parameters of the computations are available.²¹ This scatterer of index $n_i = 1.5$ lies in vacuum ($n_e = 1$). There are also four inclusions inside C_0 . The elliptical one has principal axes with half-dimensions of 1 and 0.5, its center is at $(-1, 4)$, its principal axes are rotated 45° away from the x, y axes, and it is made of infinitely conducting material. The rectangular one is placed in the region $-1.5 < x < -0.5$ and $-2 < y < 0$ and is filled with vacuum. One of the circular inclusions has its center at $(2, -1)$ and a radius of 0.5, and it is filled with vacuum. The second circular inclusion has its center at $(3, 4)$ and a radius of 0.5, and it is filled with a lossy material of optical index $0.5 + 2i$ (typical value for a metal in the optical range). This structure is illuminated with an incidence $\theta^{inc} = -45^\circ$ by a plane

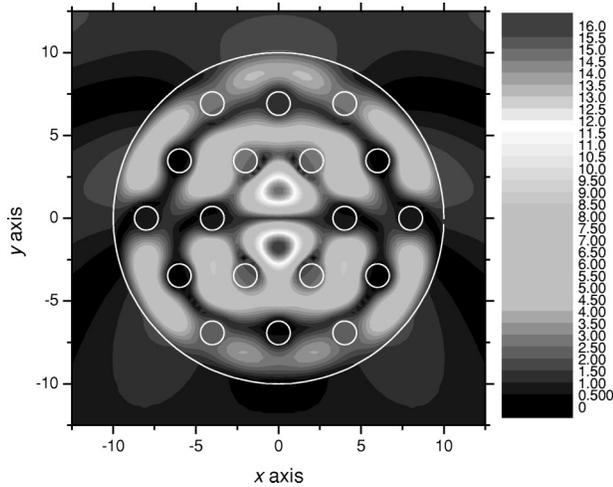


Fig. 9. Field modulus in the same conditions as in Ref. 2. The gray levels are chosen to match as closely as possible those of Ref. 2.

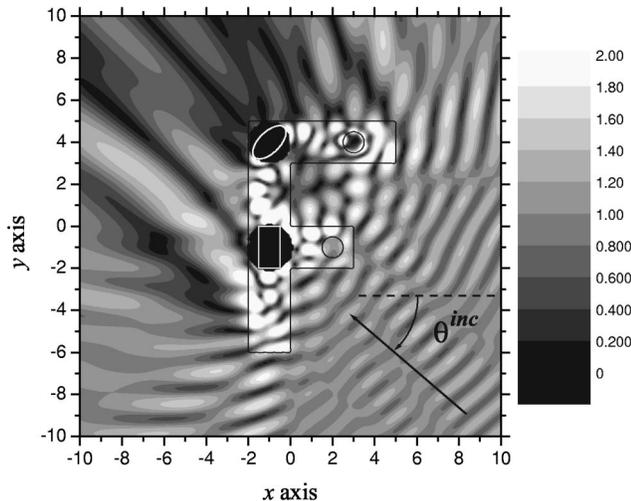


Fig. 10. Modulus of the total field for a scatterer with four inclusions.

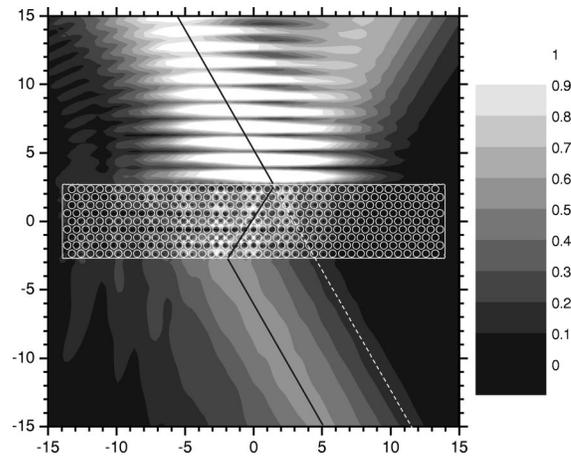


Fig. 11. Modulus of the total field. Above the slab, the slanted line shows the locus of the maximum of the Gaussian incident beam. Below the slab, it shows the locus of the maximum of the transmitted field. Above the slab, the structure of the field is due to the interference between the incident and the reflected fields.

wave of wavelength $\lambda_0 = 2$ and p polarization. In this case, the number of sources in each region Ω_e and Ω_i is taken to be $N = 500$. The total H_z field map is shown in Fig. 10. Note that the present version of our numerical code does not allow us to compute the field inside a circle that includes elliptical or rectangular bodies. This is why dark areas appear around these two inclusions.

Our last, and more practical, example illustrates the case of a dielectric slab periodically drilled with 364 circular air holes. The sides of this finite slab are defined by $x = \pm 13.944$ and $y = \pm 2.6996$ (see Fig. 11). The permittivity of the slab is $n_i^2 = 12$ and the radius of the holes is 0.294, and the holes are placed with hexagonal symmetry. The distance between the centers of two neighbor holes is 0.68. These parameters are chosen in order to exhibit a negative refraction in s polarization at the wavelength $\lambda_0 = 2.02$. The structure lies in vacuum and is illuminated by a Gaussian beam coming from the top with an incidence $\theta_0 = 30^\circ$. The exact definition of this incident field is

$$u^{inc}(x, y) = \int_{-\infty}^{+\infty} A(\alpha) \exp[i\alpha x - i\beta(\alpha)y] d\alpha, \quad (20)$$

with $\beta(\alpha)^2 = k_0^2 - \alpha^2$ and with a Gaussian amplitude:

$$A(\alpha) = \frac{W}{2\sqrt{\pi}} \exp\left[-\frac{(\alpha - \alpha_0)^2 W^2}{4}\right]. \quad (21)$$

The mean incidence $\theta_0 = 30^\circ$ of the beam is such that $\alpha_0 = k_0 \sin \theta_0$. The parameter W in Eq. (21) is directly linked to the incident beam width and has the value $W = 5$. Figure 11 shows the resulting field map, including the negative refraction inside the crystal.

7. CONCLUSION

We have presented a method that combines the advantages of the MFS and the SMM for the study of scatterers of arbitrary shape and containing arbitrary inclusions. Note that the MFS by itself could also deal with such

problems, since it is not limited to diffraction by one homogeneous body (the extension of the MFS to several bodies is straightforward). But in many cases the method presented here is more efficient, particularly when the inclusions have circular cross sections, since in that case the SMM is particularly efficient.

It is also important to note that any method able to efficiently compute the field and its derivatives inside C_0 in the presence of sources placed outside C_0 (the problem summarized in Fig. 8) could be used in place of the SMM. In other words, the MFS could be combined with various other methods by using the basic ideas described in this paper.

Let us also stress that since the method described in this paper is based on quite intuitive principles that remain true for three-dimensional problems, it can also be easily extended to this class of problem. For instance, the method could be useful for the study of problems dealing with a small number of buried objects.

Our short-term goal is to apply the method to the study of slabs made with two-dimensional photonic crystals, and in particular to the study of the negative refraction that can be observed in such structures.

Corresponding author G. Tayeb's e-mail address is gerard.tayeb@fresnel.fr.

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